





**ISO1050** 

SLLS983K - JUNE 2009 - REVISED AUGUST 2023

### ISO1050 Isolated CAN Transceiver

#### 1 Features

- Meets the requirements of ISO11898-2
- 5000-V<sub>RMS</sub> isolation (ISO1050DW)
- 2500-V<sub>RMS</sub> isolation (ISO1050DUB)
- Fail-safe outputs
- Low loop delay: 150 ns (typical), 210 ns (maximum)
- 50-kV/µs typical transient immunity
- Bus-fault protection of -27 V to 40 V
- Driver (TXD) dominant time-out function
- I/O voltage range supports 3.3 V and 5 V microprocessors
- Safety-related certifications
  - VDE approval per DIN EN IEC 60747-17 (VDE 0884-17)
  - UL 1577 approved
  - CSA approved for IEC 61010-1, IEC 60601-1
  - TUV Reinforced Insulation Approval for EN/UL/CSA 62368-1 (ISO1050DW-Only)
  - CQC reinforced insulation per GB4943.1 (ISO1050DW-only)
  - Typical 25-year life at rated working voltage (see application report SLLA197 and Life Expectancy vs Working Voltage)

### 2 Applications

- Industrial automation, control, sensors, and drive systems
- Building and climate control (HVAC) automation
- Security systems
- Transportation
- Medical
- Telecom
- CAN bus standards such as CANopen, DeviceNet, NMEA2000, ARINC825, ISO11783, CAN Kingdom, **CANaerospace**

### 3 Description

The ISO1050 is a galvanically isolated CAN transceiver that meets the specifications of the ISO11898-2 standard. The device has the logic input and output buffers separated by a silicon oxide (SiO<sub>2</sub>) insulation barrier that provides galvanic isolation of up to 5000  $V_{RMS}$  for ISO1050DW and 2500  $V_{RMS}$ for ISO1050DUB. Used in conjunction with isolated power supplies, the device prevents noise currents on a data bus or other circuits from entering the local ground and interfering with or damaging sensitive circuitry.

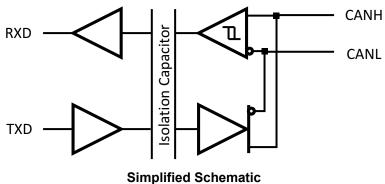
As a CAN transceiver, the device provides differential transmit capability to the bus and differential receive capability to a CAN controller at signaling rates up to 1 megabit per second (Mbps). The device is designed for operation in especially harsh environments, and it features cross-wire, overvoltage and loss of ground protection from -27 V to 40 V and overtemperature shutdown, as well as -12V to 12V common-mode range.

The ISO1050 is characterized for operation over the ambient temperature range of -55°C to 105°C.

### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
ISO1050	SOP (8)	9.50 mm × 6.57 mm
130 1030	SOIC (16)	10.30 mm × 7.50 mm

For all available packages, see the orderable addendum at the end of the datasheet.





### Table of Contents

ıa	ible of	Contents	
1 Features	1	8.2 Functional Block Diagram	19
2 Applications	1	8.3 Feature Description	
3 Description		8.4 Device Functional Modes	
4 Revision History		9 Application and Implementation	24
5 Pin Configuration and Functions		9.1 Application Information	
6 Specifications		9.2 Typical Application	
6.1 Absolute Maximum Ratings		10 Power Supply Recommendations	
6.2 ESD Ratings		10.1 General Recommendations	
6.3 Recommended Operating Conditions		11 Layout	
6.4 Thermal Information		11.1 Layout Guidelines	28
6.5 Power Ratings	<mark>7</mark>	11.2 Layout Example	
6.6 Insulation Specifications		12 Device and Documentation Support	
6.7 Safety-Related Certifications		12.1 Documentation Support	29
6.8 Safety Limiting Values		12.2 Receiving Notification of Documentation U	
6.9 Electrical Characteristics - DC Specification	10	12.3 Support Resources	29
6.10 Switching Characteristics	12	12.4 Trademarks	
6.11 Insulation Characteristics Curves	12	12.5 Electrostatic Discharge Caution	29
6.12 Typical Characteristics	13	12.6 Glossary	
7 Parameter Measurement Information		13 Mechanical, Packaging, and Orderable	
8 Detailed Description	19	Information	29
8.1 Overview			
4 Revision History NOTE: Page numbers for previous revisions ma	ny differ fi	rom page numbers in the current version.	
Changes from Povision I (September 2019)	to Bovio	ion K (August 2022)	Dago

Changes from Revision J (September 2019) to Revision K (August 2023)	Page
Changed VDE standard name to DIN EN IEC 60747-17 (VDE 0884-17), updated CSA standard 62368-1 and IEC 62368-1	
Updated the numbering format for tables, figures, and cross-references throughout the documen	
Updated Thermal Characteristics, Safety Limiting Values, and Thermal Derating Curves to provide	
accurate system-level thermal calculations	
Updated electrical and switching characteristics to match device performance	10
Changes from Revision I (September 2014) to Revision J (September 2019)	Page
Changed VDE standard name From: DIN V VDE V 0884-10 (VDE V 0884-10):2006-12 To: DIN V	/DE V
0884-11:2017-01 in Section 1	1
Deleted 'Component Acceptance Notice 5 A' from CSA bullet in Section 1	1
Changed inverting output label From: CANH To: CANL in Figure 7-10	14
Changed VDE standard name From: DIN V VDE V 0884-10 (VDE V 0884-10):2006-12 To: DIN V	
0884-11:2017-01 in INSULATION CHARACTERISTICS table	
Changed V <sub>ISO</sub> PARAMETER description From: 'ISO1050DUB - Double Protection' To: 'ISO1050I	
Protection' in INSULATION CHARACTERISTICS table	
Updated Regulatory Information in REGULATORY INFORMATION table	
Changed UL 1577 rating for ISO1050DUB From: '2500 V <sub>RMS</sub> Double Protection' To: '2500 V <sub>RMS</sub>	
Protection' in REGULATORY INFORMATION table	
Deleted UL 1577 'Double Protection' rating of 3500 V <sub>RMS</sub> for ISO1050DW in REGULATORY	19
	40
INFORMATION table	
Added SN6505x data sheet link to 'Transformer Driver for Isolated Power Supplies' in Section 12	
	<mark>29</mark>

Changes from Revision H (June 2013) to Revision I (September 2014) Added Pin Configuration and Functions section, ESD Ratings table, Feature Description section, Device

Page

Functional Modes, Application and Implementation section, Power Supply Recommendations section, Layout



section, Device and Documentation Support section, and Mechanical, Packaging, and Orderable Information section

Cł	nanges from Revision G (March 2013) to Revision H (June 2013)	Page
•	Changed title From: LIFE EXPECTANCY vs WORKING VOLTAGE (ISO1050DW To: LIFE EXPECT	
	WORKING VOLTAGE (ISO1050DUB)	
Cr	nanges from Revision F (January 2013) to Revision G (March 2013)	Page
•	Clarified clearance and creepage measurement method in ISOLATOR CHARACTERISTICS	
•	Clarified test methods for voltage ratings in INSULATION CHARACTERISTICS	
•	Changed UL Single Protection Certification pending to Single Protection in REGULATORY INFORM	/IATION
	SECTION (certificate available)	19
Cr	nanges from Revision E (December 2011) to Revision F (January 2013)	
•	Deleted ISO1050L device	
•	Deleted ISO1050LDW from Features list	
•	Deleted ISO1050LDW in first paragraph of DESCRIPTION	
•	Added the PIN FUNCTIONS section	
•	Added Maximum impulse voltage (V <sub>IMP</sub> ) specification per DIN EN IEC 60747-17 (VDE 0884-17)	8
•	Deleted ISO1050LDW from INSULATION CHARACTERISTICS	
•	Deleted ISO1050LDW from REGULATORY INFORMATION	
•	Added the FUNCTIONAL DESCRIPTION section	
•	Deleted ISO1050LDW from LIFE EXPECTANCY vs WORKING VOLTAGE	
•	Deleted 40V from the CANH and CANL input diagrams and output diagrams in the EQUIVALENT I/SCHEMATICS	
	Changed the APPLICATION INFORMATION section	
•	Changed the BUS LOADING, LENGTH AND NUMBER OF NODES section	
•	Added the CAN TERMINATION section	
 Cr	nanges from Revision D (June 2011) to Revision E (November 2011)	Page
•	Added device ISO1050L	1
•	Changed (DW Package) in the Features list to (ISO1050DW)	
•	Changed (DUB Package) in the Features list to (ISO1050DUB and ISO1050LDW)	
•	Deleted IEC 60950-1 from the CSA Approvals Feature bullet	1
•	From: IEC 60601-1 (Medical) and CSA Approvals Pending To: IEC 60601-1 (Medical) and IEC 60601-1	
•	Added Feature - 5 KVRMS Reinforced	
•	Changed DW Package to ISO105DW and DUB package to ISO1050DUB and ISO1050LDW in the paragraph of DESCRIPTION	
	Added Note 1 to the INSULATION CHARACTERISTICS table	
	Changed V <sub>IORM</sub> From: 8-DUB Package to ISO1050DUB and ISO1050LDW	
	Changed V <sub>IORM</sub> From: 16-DW to ISO1050DW	19
	Changed the V <sub>ISO</sub> Isolation voltage per UL section of the INSULATION CHARACTERISTICS table	19
•	Changed the IEC 60664-1 Ratings Table	
•	Changed the REGULATORY INFORMATION table	
•	Changed From: File Number: 220991 (Approval Pending) To: File Number: 220991	
•	Changed in note (1) 3000 to 2500 and 6000 to 5000	19
•	Changed in LIFE EXPECTANCY vs WORKING VOLTAGE (8-DUB PACKAGE TO: LIFE(ISO105	50DW and
	ISO1050LDW)	22
C.F	nanges from Revision C (July 2010) to Revision D (June 2011)	Page
•	Changed the REGULATORY INFORMATION table	



Changes from Revision B (June 2010) to Revision C (July 2010)	Page
• Changed the IEC 60747-5-2 Features bullet From: DW package Approval Pending To: VDE approved f	or
both DUB and DW packages	1
• Changed the Minimum Internal Gap value from 0.008 to 0.014 in the Isolator Characteristics table	19
Changed V <sub>IORM</sub> Specification From: 1300 To: 1200 per VDE certification	19
Changed V <sub>PR</sub> Specification From 2438 To: 2250	19
Added the Bus Loading paragraph to the Application Information section	24
Changes from Revision A (Sept 2009) to Revision B (June 2010)	Page
Added information that IEC 60747-5-2 and IEC61010-1 have been approved	1
Changed DW package from preview to production data	<mark>5</mark>
Added Insulation Characteristics and IEC 60664-1 Ratings tables	19
Added IEC file number	
Changes from Revision * (June 2009) to Revision A (Sept 2009)	Page
Added Typical 25-Year Life at Rated Working Voltage to Features	1
Added LIFE EXPECTANCY vs WORKING VOLTAGE section	<mark>22</mark>



# **5 Pin Configuration and Functions**

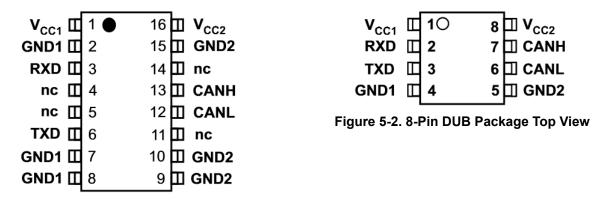


Figure 5-1. 16-Pin DW Package Top View

**Table 5-1. Pin Functions** 

	PIN		TYPE	DESCRIPTION
NAME	DW	DUB	1176	DESCRIPTION
V <sub>CC1</sub>	1	1	Supply	Digital-side supply voltage (3 to 5.5 V)
GND1	2	_	Ground	Digital-side ground connection
RXD	3	2	0	CAN receive data output (LOW for dominant and HIGH for recessive bus states)
NC	4	_	NC	No connect
NC	5	_	NC	No connect
TXD	6	3	I	CAN transmit data input (LOW for dominant and HIGH for recessive bus states)
GND1	7	4	Ground	Digital-side ground connection
GND1	8	_	Ground	Digital-side ground connection
GND2	9	5	Ground	Transceiver-side ground connection
GND2	10	_	Ground	Transceiver-side ground connection
NC	11	_	NC	No connect
CANL	12	6	I/O	Low-level CAN bus line
CANH	13	7	I/O	High-level CAN bus line
NC	14	_	NC	No connect
GND2	15	_	Ground	Transceiver-side ground connection
V <sub>CC2</sub>	16	8	Supply	Transceiver-side supply voltage (5 V)



### **6 Specifications**

### 6.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)(1) (2)

		MIN	MAX	UNIT
V <sub>CC1</sub>	Supply voltage, side 1	-0.5	6	V
V <sub>CC2</sub>	Supply voltage, side 2	-0.5	6	V
V <sub>IO</sub>	Logic input voltage range (TXD)	-0.5	V <sub>CC1</sub> +0.5 <sup>(3)</sup>	V
V <sub>BUS</sub>	Voltage on bus pins (CANH, CANL)	-27	40	V
Io	Output current on RXD pin	-15	15	mA
T <sub>J</sub>	Junction temperature	-55	150	°C
T <sub>STG</sub>	Storage temperature	-65	150	°C

- (1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values except differential I/O bus voltages are with respect to the local ground terminal (GND1 or GND2) and are peak voltage values.
- (3) Maximum voltage must not exceed 6 V

### 6.2 ESD Ratings

			VALUE	UNIT
V <sub>(ESD)</sub>	Electrostatic discharge Human body model (HBM), per ANSI/ ESDA/JEDEC JS-001	All pins <sup>(1)</sup>	±4000	V
V <sub>(ESD)</sub>	Electrostatic discharge Charged device model (CDM), per JEDEC specification JESD22-C101	All pins <sup>(2)</sup>	±1500	V
V <sub>(ESD)</sub>	Electrostatic discharge Charged machine model, ANSI/ ESDS5.2-1996	All pins <sup>(2)</sup>	±200	V

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### **6.3 Recommended Operating Conditions**

		MIN	TYP MAX	UNIT
V <sub>CC1</sub>	Supply Voltage, Side 1	3	5.5	V
V <sub>CC2</sub>	Supply Voltage, Side 2	4.75	5.25	V
V <sub>I</sub> or V <sub>IC</sub>	Voltage at bus pins (separately or common mode)	-12	12	V
V <sub>IH</sub>	High-level input voltage (TXD)	2	5.25	V
V <sub>IL</sub>	Low-level input voltage (TXD)	0	0.8	V
V <sub>ID</sub>	Differential input voltage	-7	7	V
I <sub>OH</sub>	High-Level Output current, Driver	-70		mA
I <sub>OH</sub>	High-Level Output current, Receiver	-4		mA
I <sub>OL</sub>	Low-level output current, Driver		70	mA
I <sub>OL</sub>	Low-level output current, Receiver		4	mA
T <sub>A</sub>	Operating ambient temperature	-55	105	°C
TJ	Junction temperature	-55	125	°C
T <sub>Jshutdown</sub>	Thermal shutdown temperature		190	°C

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### **6.4 Thermal Information**

		ISC	ISO1050		
	THERMAL METRIC(1)	DW	DUB	UNIT	
		16 PINS	8 PINS		
R <sub>⊙JA</sub>	Junction-to-ambient thermal resistance	76.4	84.3	°C/W	
R <sub>OJC(top)</sub>	Junction-to-case (top) thermal resistance	41	63.2	°C/W	
R <sub>⊝JB</sub>	Junction-to-board thermal resistance	47.7	43	°C/W	
$\Psi_{JT}$	Junction-to-top characterization parameter	17.2	27.4	°C/W	
$\Psi_{JB}$	Junction-to-board characterization parameter	38.2	42.7	°C/W	
R <sub>OJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	n/a	n/a	°C/W	

<sup>(1)</sup> For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

# **6.5 Power Ratings**

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
P <sub>D</sub>	Maximum power dissipation (both sides)	$V_{CC1}$ = $V_{CC2}$ = 5.25 V, $T_J$ = 150°C, $R_L$ = 60 $\Omega$ TXD with 5V, 500kHz 50% duty square wave			200	mW
P <sub>D1</sub>	Maximum power dissipation (side-1)	$V_{CC1}$ = $V_{CC2}$ = 5.25 V, $T_J$ = 150°C, $R_L$ = 60 $\Omega$ , TXD with 5V, 500kHz 50% duty square wave			25	mW
P <sub>D2</sub>	Maximum power dissipation (side-2)	$V_{CC1}$ = $V_{CC2}$ = 5.25 V, $T_J$ = 150°C, $R_L$ = 60 $\Omega$ TXD with 5V, 500kHz 50% duty square wave			175	mW



### 6.6 Insulation Specifications

	DADAMETER	TEGT COMPLETIONS	SPECIFICATIONS		LIND
PARAMETER		TEST CONDITIONS	DUB-8	DW-16	UNIT
IEC 6066	64-1				
CLR	External clearance <sup>(1)</sup>	Side 1 to side 2 distance through air	>6.1	>8	mm
CPG	External Creepage <sup>(1)</sup>	Side 1 to side 2 distance across package surface	>6.8	>8	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	>17	>17	μm
СТІ	Comparative tracking index	IEC 60112; UL 746A	>600	>600	V
	Material Group	According to IEC 60664-1	I	I	
		Rated mains voltage ≤ 150 V <sub>RMS</sub>	I-IV	I-IV	
	Overveltage estagen.	Rated mains voltage ≤ 300 V <sub>RMS</sub>	1-111	I-III	
	Overvoltage category	Rated mains voltage ≤ 600 V <sub>RMS</sub>	n/a	I-II	
		Rated mains voltage ≤ 848 V <sub>RMS</sub>	n/a	I	
DIN V V	DE V 0884-11:2017-01 <sup>(2)</sup>		•	•	
V <sub>IORM</sub>	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	560	1200	V <sub>PK</sub>
$V_{IOWM}$	Maximum isolation working voltage	AC voltage (sine wave); time-dependent dielectric breakdown (TDDB) test;	395	848	V <sub>RM</sub> s
		DC voltage	560	1200	V <sub>DC</sub>
V <sub>IOTM</sub>	Maximum transient isolation voltage	$V_{TEST} = V_{IOTM}$ , t = 60 s (qualification); $V_{TEST}$ = 1.2 × $V_{IOTM}$ , t = 1 s (100% production)	4000	4000	V <sub>PK</sub>
$V_{IMP}$	Maximum impulse voltage <sup>(3)</sup>	Tested in air, 1.2/50-us waveform per IEC 62368-1	3076	3076	V <sub>PK</sub>
V <sub>IOSM</sub>	Maximum surge isolation voltage <sup>(3)</sup>	V <sub>IOSM</sub> ≥ 1.3 x V <sub>IMP</sub> ; Tested in oil (qualification test), 1.2/50-µs waveform per IEC 62368-1	4000	4000	V <sub>PK</sub>
		Method a: After I/O safety test subgroup 2/3, $V_{ini} = V_{IOTM}$ , $t_{ini} = 60$ s; $V_{pd(m)} = 1.2 \times V_{IORM}$ , $t_m = 10$ s	≤ 5	≤ 5	
$q_{pd}$	Apparent charge <sup>(4)</sup>	Method a: After environmental tests subgroup 1, $V_{ini} = V_{IOTM}$ , $t_{ini} = 60 \text{ s}$ ; $V_{pd(m)} = 1.6 \times V_{IORM}$ , $t_m = 10 \text{ s}$	≤ 5	≤ 5	pC
		Method b: At routine test (100% production) $V_{ini} = 1.2 \times V_{IOTM}$ , $t_{ini} = 1 \text{ s}$ ; $V_{pd(m)} = 1.5 \times V_{IORM}$ , $t_m = 1 \text{ s}$	≤ 5	≤ 5	
C <sub>IO</sub>	Barrier capacitance, input to output <sup>(5)</sup>	V <sub>IO</sub> = 0.4 × sin (2 πft), f = 1 MHz	1	1	pF
		V <sub>IO</sub> = 500 V, T <sub>A</sub> = 25°C	> 10 <sup>12</sup>	> 10 <sup>12</sup>	
R <sub>IO</sub>	Insulation resistance, input to output <sup>(5)</sup>	V <sub>IO</sub> = 500 V, 100°C ≤ T <sub>A</sub> ≤ 150°C	> 10 <sup>11</sup>	> 10 <sup>11</sup>	Ω
		V <sub>IO</sub> = 500 V at T <sub>S</sub> = 150°C	> 10 <sup>9</sup>	> 10 <sup>9</sup>	
	Pollution degree		2	2	
	Climatic category		40/125/ 21	40/125/ 21	
UL 1577			•	•	•
V <sub>ISO</sub>	Withstand isolation voltage	$V_{TEST} = V_{ISO}$ , t = 60 s (qualification); $V_{TEST} = 1.2 \times V_{ISO}$ , t = 1 s (100% production)	2500	4243	V <sub>RMS</sub>

<sup>(1)</sup> Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves, ribs, or both on a printed circuit board are used to help increase these specifications.

<sup>(2)</sup> ISO1044 is suitable for basic electrical insulation only within the safety ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.

<sup>(3)</sup> Testing is carried out in air or oil to determine the intrinsic surge immunity of the isolation barrier.

<sup>(4)</sup> Apparent charge is electrical discharge caused by a partial discharge (pd).

(5) All pins on each side of the barrier tied together creating a two-pin device.

### 6.7 Safety-Related Certifications

VDE	CSA	UL	CQC	TUV
Plan to certify according to DIN V VDE V 0884-11:2017- 01	Plan to certify according to IEC 60950-1, IEC 62368-1	Plan to certify according to UL 1577 Component Recognition Program	Plan to certify according to GB4943.1-2011	Plan to certify according to EN 61010-1:2010 (3rd Ed) and EN 60950- 1:2006/A11:2009/A1:2010/ A12:2011/A2:2013
Maximum transient isolation voltage, 4242 V <sub>PK</sub> ; Maximum repetitive peak isolation voltage, 637 V <sub>PK</sub> ; Maximum surge isolation voltage, 5000 V <sub>PK</sub>	400 V <sub>RMS</sub> basic insulation working voltage per CSA 60950-1-07+A1+A2 and IEC 60950-1 2nd Ed., for pollution degree 2, material group I	Single protection, 3000 V <sub>RMS</sub>	Basic Insulation, Altitude ≤ 5000 m, Tropical Climate, 400 V <sub>RMS</sub> maximum working voltage	EN 61010-1:2010 (3rd Ed) 300 V <sub>RMS</sub> working voltage 
Certificate planned	Certificate planned	Certificate planned	Certificate planned	Certificate planned

### 6.8 Safety Limiting Values

Safety limiting<sup>(1)</sup> intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
DUB-8	PACKAGE				-	
Is	Safety input, output, or supply current	$R_{\theta JA} = 84.3$ °C/W, $V_I = 5.5$ V, $T_J = 150$ °C, $T_A = 25$ °C. see Figure 6-1			269	mA
Is	Safety input, output, or supply current	$R_{\theta JA} = 84.3$ °C/W, $V_I = 3.6$ V, $T_J = 150$ °C, $T_A = 25$ °C. Figure 6-1			411	mA
T <sub>S</sub>	Maximum safety temperature				150	°C
DW-16	Package				•	
Is	Safety input, output, or supply current	$R_{\theta JA} = 76.4$ °C/W, $V_I = 5.5$ V, $T_J = 150$ °C, $T_A = 25$ °C.			297	mA
Is	Safety input, output, or supply current	$R_{\theta JA} = 76.4$ °C/W, $V_I = 3.6$ V, $T_J = 150$ °C, $T_A = 25$ °C.			454	mA
T <sub>S</sub>	Maximum safety temperature				150	°C

<sup>(1)</sup> The maximum safety temperature, T<sub>S</sub>, has the same value as the maximum junction temperature, T<sub>J</sub>, specified for the device. The I<sub>S</sub> and P<sub>S</sub> parameters represent the safety current and safety power respectively. The maximum limits of I<sub>S</sub> and P<sub>S</sub> should not be exceeded. These limits vary with the ambient temperature, T<sub>A</sub>.

The junction-to-air thermal resistance, R<sub>θJA</sub>, in the table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:

 $T_J = T_A + R_{\theta JA} \times P$ , where P is the power dissipated in the device.

 $T_{J(max)} = T_S = T_A + R_{\theta JA} \times P_S$ , where  $T_{J(max)}$  is the maximum allowed junction temperature.

 $P_S = I_S \times V_I$ , where  $V_I$  is the maximum input voltage.



# 6.9 Electrical Characteristics - DC Specification

Typical specifications are at  $V_{CC1}$  = 3.3V,  $V_{CC2}$  = 5V, Min/Max are over recommended operating conditions (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SUPPLY	CHARACTERISTICS					
I <sub>CC1</sub>	Supply current Side 1	V <sub>I</sub> = 0 V or V <sub>CC1</sub> , V <sub>CC1</sub> = 3.3 V		1.8	3.6	mA
I <sub>CC1</sub>	Supply current Side 1	V <sub>I</sub> = 0 V or V <sub>CC1</sub> , V <sub>CC1</sub> = 5.0 V		2.3	3.6	mA
I <sub>CC2</sub>	Supply current Side 2	$V_{l} = 0 \text{ V}$ , bus dominant, $R_{L} = 60 \Omega$		52	73	mA
I <sub>CC2</sub>	Supply current Side 2	V <sub>I</sub> = V <sub>CC1</sub>		8	12	mA
	ELECTRICAL CHARACTERISTICS					
.,	Bus output voltage(Dominant), CANH	See Figure 7-1 and Figure 7-2, $V_I$ = 0 V, $R_L$ = 60 $\Omega$	2.9	3.5	4.5	V
$V_{O(DOM)}$	Bus output voltage(Dominant), CANL	See Figure 7-1 and Figure 7-2, $V_I$ = 0 V, $R_L$ = 60 $\Omega$	0.8	1.2	1.78	V
V <sub>O(REC)</sub>	Bus output voltage(recessive), CANH and CANL	See Figure 7-1 and Figure 7-2, $V_I$ = 2 V, $R_L$ = 60 $\Omega$	2.0	2.3	3.0	V
.,	Differential automotoral accordance (description)	See Figure 7-1 and Figure 7-2, $V_I$ = 0 V, $R_L$ = 60 $\Omega$	1.5		3.0	V
V <sub>OD(DOM)</sub>	Differential output voltage(dominant)	See Figure 7-1 and Figure 7-2, $V_I$ = 0 V, $R_L$ = 45 $\Omega$ , VCC > 4.8 V	1.4		3.0	V
V <sub>OD(REC)</sub>	Differential output voltage(recessive)	See Figure 7-1 and Figure 7-2, $V_l$ = 3 $V_r$ , $R_L$ = 60 $\Omega$	-120.0		12.0	mV
()		V <sub>I</sub> = 3 V, No Load	-500.0		50.0	mV
V <sub>OC(DOM)</sub>	Common-mode output voltage (Dominant)	See Figure 7-8	2	2.3	3.0	V
V <sub>OC(pp)</sub>	Peak-to-peak common-mode output voltage	See Figure 7-8		0.3		V
I <sub>IH</sub>	High level input leakage current	V <sub>I</sub> = 2 V			5	uA
I <sub>IL</sub>	Low level input leakage current	V <sub>I</sub> = 0.8 V	-5			uA
I <sub>O(off)</sub>	Power-off TXD leakage current	V <sub>CC1</sub> , V <sub>CC2</sub> at 0 V, TXD = 5 V			10	uA
		See Figure 7-11, CANH = -12 V, CANL Open	-105	-72		mA
	Short circuit current steady state output	See Figure 7-11, VCANH = 12 V, CANL Open		0.36	6.2	mA
I <sub>OS(ss)</sub>	current, dominant	See Figure 7-11, VCANL =-12 V, CANH Open	-1	-0.5		mA
		See Figure 7-11, VCANL = 12 V, CANH Open		71	105	mA
СМТІ	Common-mode transient immunity	See Figure 7-13, V <sub>I</sub> = V <sub>CC</sub> or 0 V	25	50		kV/us
RECEIVE	R ELECTRICAL CHARACTERISTICS					
V <sub>IT+</sub>	Positive-going bus input threshold voltage			750	900.0	mV
V <sub>IT-</sub>	Negative-going bus input threshold voltage	See Table 1	500.0	650		mV
V <sub>HYS</sub>	Hysteresis voltage for differential input threshold			150		mV
Vau	High level output voltage with Vcc = 5 V	I <sub>O</sub> = -4 mA, See Figure 7-6	Vcc - 0.8	4.6		V
V <sub>OH</sub>	riigir ievei output voitage witti vcc – 3 v	I <sub>O</sub> = -20 uA, See Figure 7-6	Vcc - 0.1	5		V
	High level output voltage with Vcc1 = 3.3	I <sub>O</sub> = 4 mA, See Figure 7-6	Vcc - 0.8	3.1		V
V <sub>OH</sub>	V	I <sub>O</sub> = 20 uA, See Figure 7-6	Vcc - 0.1	3.3		V
\/	Low level output voltage	I <sub>O</sub> = 4 mA, See Figure 7-6		0.2	0.4	V
$V_{OL}$	Low level output voltage	I <sub>O</sub> = 20 uA, See Figure 7-6		0	0.1	V

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Typical specifications are at  $V_{CC1}$  = 3.3V,  $V_{CC2}$  = 5V, Min/Max are over recommended operating conditions (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Cı	Input capacitance to ground (CANH or CANL)	TXD = 3 V, V <sub>I</sub> = 0.4 sin (4e6pi*t) + 2.5 V		12		pF
C <sub>ID</sub>	Differential input capacitance	TXD = 3 V, V <sub>I</sub> = 0.4 sin (4e6pi*t)		8		pF
R <sub>ID</sub>	Differential input resistance	TXD = 3 V	40		90	kΩ
R <sub>IN</sub>	Input resistance (CANH or CANL)	TXD = 3 V	20		45	kΩ
R <sub>IN(M)</sub>	Input resistance matching: (1 - R <sub>IN(CANH)</sub> /R <sub>IN(CANL)</sub> ) x 100%	V <sub>CANH</sub> = V <sub>CANL</sub>	-3		3	%
CMTI	Common-mode transient immunity	See Figure 7-13, V <sub>I</sub> = V <sub>CC</sub> or 0 V	25	50		kV/us
THERMA	AL SHUTDOWN				'	

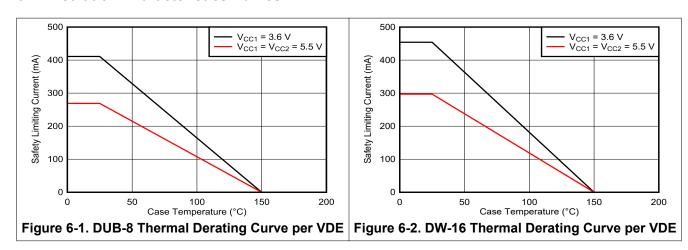


### **6.10 Switching Characteristics**

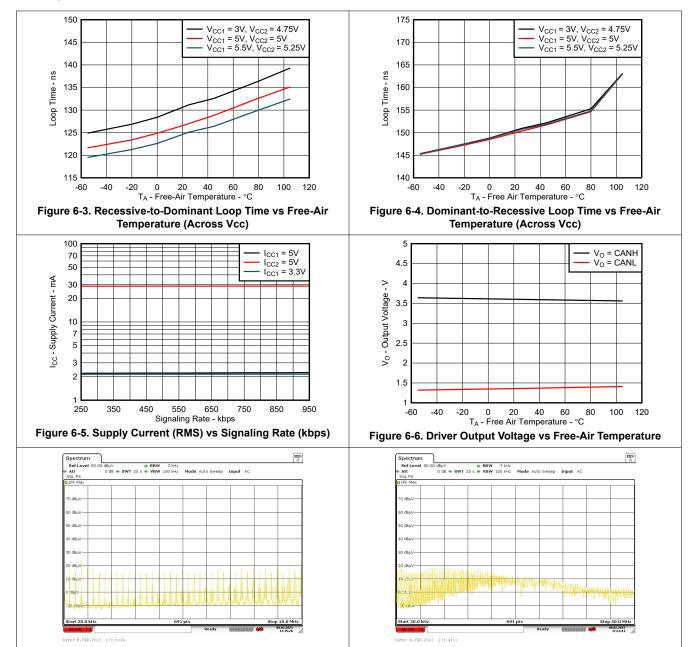
Typical specifications are at  $V_{CC1}$  = 3.3V,  $V_{CC2}$  = 5V, Min/Max are over recommended operating conditions (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
DEVICE S	SWITCHING CHARACTERISTICS		•			
t <sub>PROP(LOO</sub>	Total loop delay, driver input TXD to receiver RXD, recessive to dominant	See Figure 7-9	100	150	210	ns
t <sub>PROP(LOO</sub>	Total loop delay, driver input TXD to receiver RXD, dominant to recessive	See Figure 7-9	112	150	210	ns
DRIVER S	SWITCHING CHARACTERISTICS		•			
t <sub>PLH</sub>	Propagation delay time, recessive-to-dominant output			74	110	
t <sub>PHL</sub>	Propagation delay time, dominant-to-recessive output	See Figure 7-4		82	110	ns
t <sub>R</sub>	Differential output signal rise time			20	50	
t <sub>F</sub>	Differential output signal fall time			52	63	
t <sub>TXD_DTO</sub>	Dominant time out	C <sub>L</sub> = 100 pF, See Figure 7-10	1.2		4	ms
RECEIVE	R SWITCHING CHARACTERISTICS			-		
t <sub>PLH</sub>	Propagation delay time, low-to-high-level output		66	90	130	ns
t <sub>PHL</sub>	Propagation delay time, high-to-low-level output	TXD at 3 V, See Figure 7-6	51	80	105	ns
t <sub>R</sub>	Output signal rise time(RXD)			3	6	ns
t <sub>F</sub>	Output signal fall time(RXD)			3	6	ns
t <sub>fs</sub>	Fail-Safe output delay time from bus-side power loss	VCC1 at 5 V, See Figure 7-12		6		us

### **6.11 Insulation Characteristics Curves**



### 6.12 Typical Characteristics



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Figure 6-7. Emissions Spectrum to 10 MHz

Figure 6-8. Emissions Spectrum to 50 MHz



### 7 Parameter Measurement Information

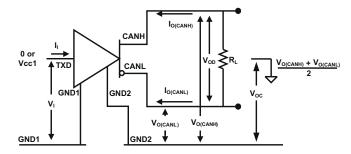


Figure 7-1. Driver Voltage, Current and Test Definitions

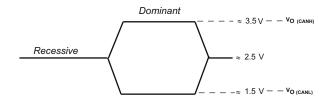


Figure 7-2. Bus Logic State Voltage Definitions

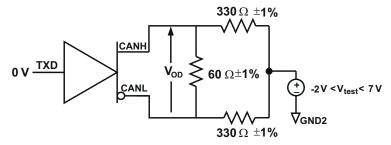
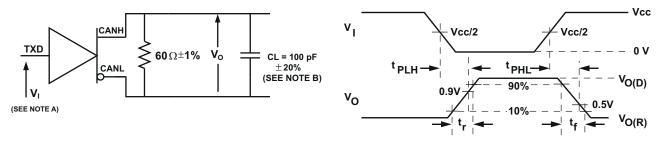


Figure 7-3. Driver V<sub>OD</sub> With Common-Mode Loading Test Circuit



- A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  125 kHz, 50% duty cycle,  $t_r \leq$  6 ns,  $t_f \leq$  6 ns,  $Z_O =$  50  $\Omega$ .
- B. C<sub>L</sub> includes instrumentation and fixture capacitance within ±20%.

Figure 7-4. Driver Test Circuit and Voltage Waveforms

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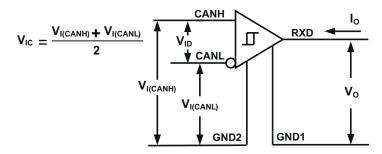
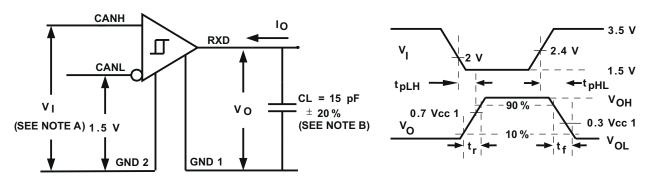


Figure 7-5. Receiver Voltage and Current Definitions

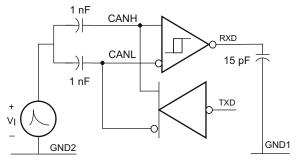


- A. The input pulse is supplied by a generator having the following characteristics: PRR ≤ 125 kHz, 50% duty cycle, t<sub>r</sub> ≤ 6 ns, t<sub>f</sub> ≤ 6 ns, Z<sub>O</sub> = 50 O
- B.  $C_L$  includes instrumentation and fixture capacitance within  $\pm 20\%$ .

Figure 7-6. Receiver Test Circuit and Voltage Waveforms

Table 7-1. Differential Input Voltage Threshold Test

	INPUT			OUTPUT
V <sub>CANH</sub>	V <sub>CANL</sub>	V <sub>ID</sub>		R
–11.1 V	–12 V	900 mV	L	
12 V	11.1 V	900 mV	L	
–6 V	–12 V	6 V	L	- V <sub>OL</sub>
12 V	6 V	6 V	L	
–11.5 V	–12 V	500 mV	Н	
12 V	11.5 V	500 mV	Н	
–12 V	−6 V	−6 V	Н	V <sub>OH</sub>
6 V	12 V	−6 V	Н	
Open	Open	X	Н	



The waveforms of the applied transients are in accordance with ISO 7637 part 1, test pulses 1, 2, 3a, and 3b.

Figure 7-7. Transient Overvoltage Test Circuit



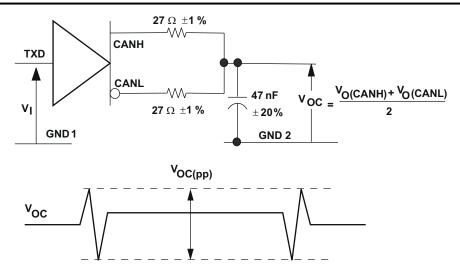


Figure 7-8. Peak-to-Peak Output Voltage Test Circuit and Waveform

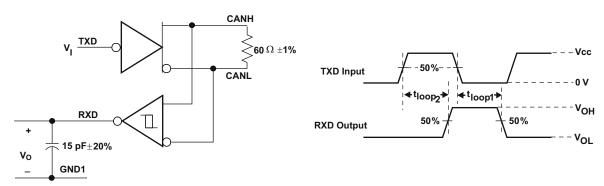
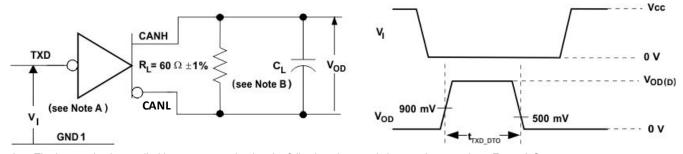


Figure 7-9. t<sub>LOOP</sub> Test Circuit and Voltage Waveforms



- A. The input pulse is supplied by a generator having the following characteristics:  $t_r \le 6$  ns,  $t_f \le 6$  ns,  $t_O = 50$   $\Omega$ .
- B.  $C_L$  includes instrumentation and fixture capacitance within  $\pm 20\%$ .

Figure 7-10. Dominant Time-out Test Circuit and Voltage Waveforms

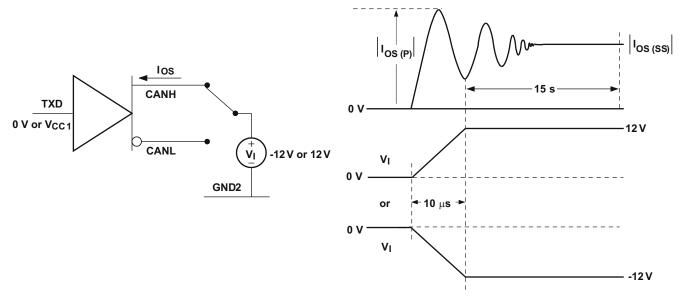


Figure 7-11. Driver Short-Circuit Current Test Circuit and Waveforms

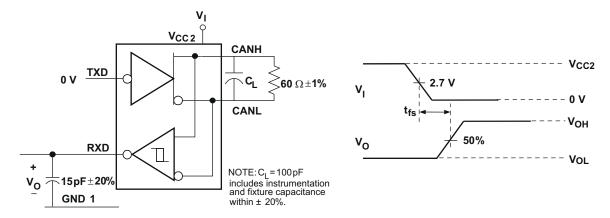


Figure 7-12. Fail-Safe Delay Time Test Circuit and Voltage Waveforms



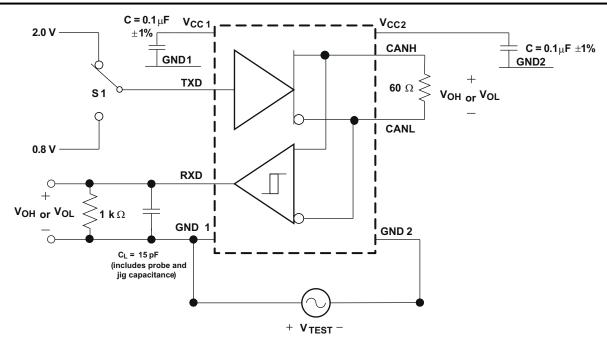


Figure 7-13. Common-Mode Transient Immunity Test Circuit

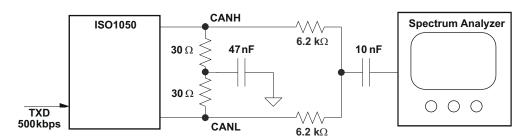


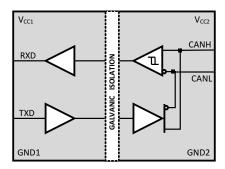
Figure 7-14. Electromagnetic Emissions Measurement Setup

### **8 Detailed Description**

### 8.1 Overview

The ISO1050 is a digitally isolated CAN transceiver with a typical transient immunity of 50 kV/µs. The device can operate from 3.3-V supply on side 1 and 5-V supply on side 2. This is of particular advantage for applications operating in harsh industrial environments because the 3.3 V on side 1 enables the connection to low-volt microcontrollers for power preservation, whereas the 5 V on side 2 maintains a high signal-to-noise ratio of the bus signals.

### 8.2 Functional Block Diagram



### 8.3 Feature Description

#### 8.3.1 CAN Bus States

The CAN bus has two states during operation: *dominant* and *recessive*. A dominant bus state, equivalent to logic low, is when the bus is driven differentially by a driver. A recessive bus state is when the bus is biased to a common mode of  $V_{CC}$  / 2 through the high-resistance internal input resistors of the receiver, equivalent to a logic high. The host microprocessor of the CAN node will use the TXD pin to drive the bus and will receive data from the bus on the RXD pin. See Figure 8-1 and Figure 8-2.

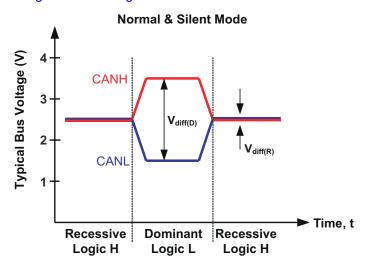


Figure 8-1. Bus States (Physical Bit Representation)



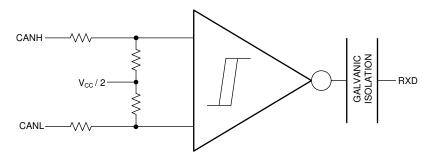


Figure 8-2. Simplified Recessive Common Mode Bias and Receiver

#### 8.3.2 Digital Inputs and Outputs

### TXD (Input) and RXD (Output):

 $V_{CC1}$  for the isolated digital input and output side of the device maybe supplied by a 3.3-V or 5-V supply and thus the digital inputs and outputs are 3.3-V and 5-V compatible.

#### Note

TXD is very weakly internally pulled up to  $V_{CC1}$ . An external pullup resistor should be used to make sure that TXD is biased to recessive (high) level to avoid issues on the bus if the microprocessor doesn't control the pin and TXD floats. TXD pullup strength and CAN bit timing require special consideration when the device is used with an open-drain TXD output on the CAN controller of the microprocessor. An adequate external pullup resistor must be used to ensure that the TXD output of the microprocessor maintains adequate bit timing input to the input on the transceiver.

#### 8.3.3 Protection Features

#### 8.3.3.1 TXD Dominant Time-Out (DTO)

TXD DTO circuit prevents the local node from blocking network communication in the event of a hardware or software failure where TXD is held dominant longer than the time-out period  $t_{TXD\_DTO}$ . The TXD DTO circuit timer starts on a falling edge on TXD. The TXD DTO circuit disables the CAN bus driver if no rising edge is seen before the time-out period expires. This frees the bus for communication between other nodes on the network. The CAN driver is re-activated when a recessive signal is seen on the TXD pin, thus clearing the TXD DTO condition. The receiver and RXD pin still reflect the CAN bus, and the bus pins are biased to recessive level during a TXD dominant time-out.

#### **Note**

The minimum dominant TXD time allowed by the TXD DTO circuit limits the minimum possible transmitted data rate of the device. The CAN protocol allows a maximum of eleven successive dominant bits (on TXD) for the worst case, where five successive dominant bits are followed immediately by an error frame. This, along with the  $t_{TXD\_DTO}$  minimum, limits the minimum data rate. Calculate the minimum transmitted data rate by: Minimum Data Rate = 11 /  $t_{TXD\_DTO}$ .

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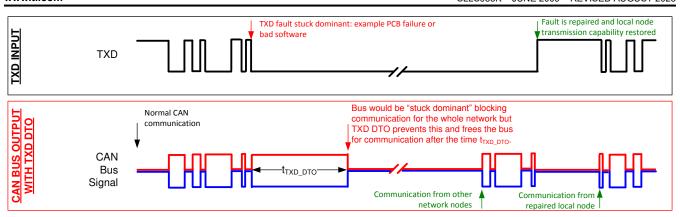


Figure 8-3. Example Timing Diagram for Devices With TXD DTO

#### 8.3.3.2 Thermal Shutdown

If the junction temperature of the device exceeds the thermal shut down threshold the device turns off the CAN driver circuits thus blocking the TXD to bus transmission path. The shutdown condition is cleared when the junction temperature drops below the thermal shutdown temperature of the device. If the fault condition is still present, the temperature may rise again and the device would enter thermal shut down again. Prolonged operation with thermal shutdown conditions may affect device reliability.

#### Note

During thermal shutdown the CAN bus drivers turn off; thus no transmission is possible from TXD to the bus. The CAN bus pins are biased to recessive level during a thermal shutdown, and the receiver to RXD path remains operational.

#### 8.3.3.3 Undervoltage Lockout and Fail-Safe

The supply pins have undervoltage detection that places the device in protected or fail-safe mode. This protects the bus during an undervoltage event on  $V_{CC1}$  or  $V_{CC2}$  supply pins. If the bus-side power supply  $V_{CC2}$  is lower than about 4 V, the power shutdown circuits in the ISO1050 will disable the transceiver to prevent false transmissions due to an unstable supply. If  $V_{CC1}$  is still active when this occurs, the receiver output (RXD) will go to a fail-safe HIGH (recessive) value in about 6 microseconds.

Table 8-1. Undervoltage Lockout and Fail-Safe

V <sub>CC</sub> 1	V <sub>CC</sub> 2	DEVICE STATE	BUS OUTPUT	RXD
GOOD	GOOD	Functional	Per Device State and TXD	Mirrors Bus
BAD	GOOD	Protected	Recessive	High Impedance (3-state)
GOOD	BAD	Protected	High Impedance	Recessive (Fail-Safe High)

#### Note

After an undervoltage condition is cleared and the supplies have returned to valid levels, the device typically resumes normal operation in 300  $\mu s$ 

### 8.3.3.4 Floating Pins

Pullups and pulldowns should be used on critical pins to place the device into known states if the pins float. The TXD pin should be pulled up through a resistor to  $V_{CC1}$  to force a recessive input level if the microprocessor output to the pin floats.

#### 8.3.3.5 CAN Bus Short-Circuit Current Limiting

The device has several protection features that limit the short-circuit current when a CAN bus line is shorted. These include driver current limiting (dominant and recessive). The device has TXD dominant state time out to prevent permanent higher short-circuit current of the dominant state during a system fault. During CAN communication the bus switches between dominant and recessive states with the data and control fields bits, thus the short-circuit current may be viewed either as the instantaneous current during each bus state, or as a DC average current. For system current (power supply) and power considerations in the termination resistors and common-mode choke ratings, use the average short-circuit current. Determine the ratio of dominant and recessive bits by the data in the CAN frame plus the following factors of the protocol and PHY that force either recessive or dominant at certain times:

- · Control fields with set bits
- Bit-stuffing
- · Interframe space
- TXD dominant time-out (fault case limiting)

These ensure a minimum recessive amount of time on the bus even if the data field contains a high percentage of dominant bits.

#### Note

The short-circuit current of the bus depends on the ratio of recessive to dominant bits and their respective short-circuit currents. The average short-circuit current may be calculated with the following formula:

 $I_{OS(AVG)}$  = %Transmit × [(%REC\_Bits ×  $I_{OS(SS)\_REC}$ ) + (%DOM\_Bits ×  $I_{OS(SS)\_DOM}$ )] + [%Receive ×  $I_{OS(SS)\_REC}$ ]

#### Where

- I<sub>OS(AVG)</sub> is the average short-circuit current.
- %Transmit is the percentage the node is transmitting CAN messages.
- %Receive is the percentage the node is receiving CAN messages.
- %REC Bits is the percentage of recessive bits in the transmitted CAN messages.
- %DOM Bits is the percentage of dominant bits in the transmitted CAN messages.
- I<sub>OS(SS)</sub> <sub>REC</sub> is the recessive steady state short-circuit current.
- I<sub>OS(SS)</sub> DOM is the dominant steady state short-circuit current.

#### Note

Consider the short-circuit current and possible fault cases of the network when sizing the power ratings of the termination resistance and other network components.

#### 8.4 Device Functional Modes

**Table 8-2. Driver Function Table** 

INPUT	OUT	PUTS	DRIVEN BUS STATE
TXD <sup>(1)</sup>	CANH <sup>(1)</sup>	CANL <sup>(1)</sup>	DRIVEN BOS STATE
L	Н	L	Dominant
Н	Z	Z	Recessive

(1) H = high level, L = low level, Z = common mode (recessive) bias to V<sub>CC</sub> / 2. See Figure 8-1 and Figure 8-2 for bus state and common mode bias information.

Table	8-3	Rec	eiver	Fun	ction	Table
Iable	U-J.	NEC	CIVCI	I UII	GHUH	Iable

DEVICE MODE	CAN DIFFERENTIAL INPUTS  V <sub>ID</sub> = V <sub>CANH</sub> - V <sub>CANL</sub>	BUS STATE	RXD PIN <sup>(1)</sup>
	V <sub>ID</sub> ≥ 0.9 V	Dominant	L
Normal or Silant	0.5 V < V <sub>ID</sub> < 0.9 V	?	?
Normal or Silent	V <sub>ID</sub> ≤ 0.5 V	Recessive	Н
	Open (V <sub>ID</sub> ≈ 0 V)	Open	Н

(1) H = high level, L = low level, ? = indeterminate.

**Table 8-4. Function Table** 

DRIVER				RECEIVER				
INPUTS	OUT	PUTS	BUS STATE	DIFFERENTIAL INPUTS	OUTPUT	BUS STATE		
TXD	CANH	CANL	BUSSIAIE	V <sub>ID</sub> = CANH-CANL	RXD	BUS STATE		
L(1)	Н	L	DOMINANT	V <sub>ID</sub> ≥ 0.9 V	L	DOMINANT		
Н	Z	Z	RECESSIVE	0.5 V < V <sub>ID</sub> < 0.9 V	?	?		
Open	Z	Z	RECESSIVE	V <sub>ID</sub> ≤ 0.5 V	Н	RECESSIVE		
X	Z	Z	RECESSIVE	Open	Н	RECESSIVE		

(1) Logic low pulses to prevent dominant time-out.

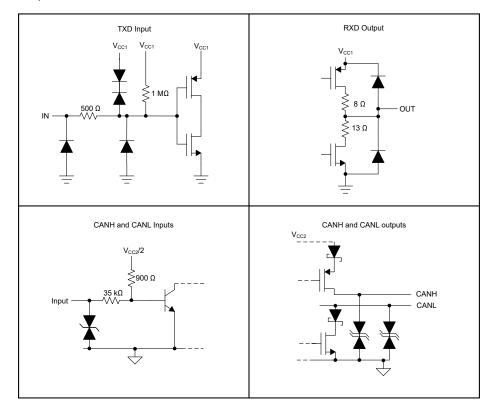


Figure 8-4. Equivalent I/O Schematics



### 9 Application and Implementation

#### Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

### 9.1 Application Information

ISO1050 can be used with other components from TI such as a microcontroller, a transformer driver, and a linear voltage regulator to form a fully isolated CAN interface.

### 9.2 Typical Application

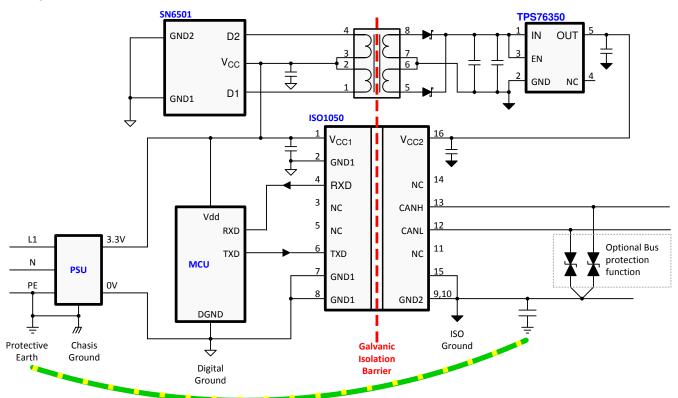


Figure 9-1. Application Circuit

#### 9.2.1 Design Requirements

Unlike optocoupler-based solution, which needs several external components to improve performance, provide bias, or limit current, ISO1050 only needs two external bypass capacitors to operate.

### 9.2.2 Detailed Design Procedure

### 9.2.2.1 Bus Loading, Length and Number of Nodes

The ISO11898 Standard specifies a maximum bus length of 40 m and maximum stub length of 0.3 m with a maximum of 30 nodes. However, with careful design, users can have longer cables, longer stub lengths, and many more nodes to a bus. A high number of nodes requires a transceiver with high input impedance such as the ISO1050.

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Many CAN organizations and standards have scaled the use of CAN for applications outside the original ISO11898 standard. They have made system level trade offs for data rate, cable length, and parasitic loading of the bus. Examples of some of these specifications are ARINC825, CANopen, CAN Kingdom, DeviceNet and NMEA200.

A CAN network design is a series of tradeoffs, but these devices operate over wide -12-V to 12-V common-mode range. In ISO11898-2 the driver differential output is specified with a  $60\text{-}\Omega$  load (the two  $120\text{-}\Omega$  termination resistors in parallel) and the differential output must be greater than 1.5 V. The ISO1050 is specified to meet the 1.5-V requirement with a  $60\text{-}\Omega$  load, and additionally specified with a differential output of 1.4 V with a  $45\text{-}\Omega$  load. The differential input resistance of the ISO1050 is a minimum of  $30\text{ k}\Omega$ . If 167 ISO1050 transceivers are in parallel on a bus, this is equivalent to a  $180\text{-}\Omega$  differential load. That transceiver load of 180  $\Omega$  in parallel with the 60  $\Omega$  gives a total 45  $\Omega$ . Therefore, the ISO1050 theoretically supports over 167 transceivers on a single bus segment with margin to the 1.2-V minimum differential input at each node. However for CAN network design margin must be given for signal loss across the system and cabling, parasitic loadings, network imbalances, ground offsets and signal integrity thus a practical maximum number of nodes is typically much lower. Bus length may also be extended beyond the original ISO11898 standard of 40 m by careful system design and data rate tradeoffs. For example, CAN open network design guidelines allow the network to be up to 1km with changes in the termination resistance, cabling, less than 64 nodes and significantly lowered data rate.

This flexibility in CAN network design is one of the key strengths of the various extensions and additional standards that have been built on the original ISO11898 CAN standard. In using this flexibility comes the responsibility of good network design.

#### 9.2.2.2 CAN Termination

The ISO11898 standard specifies the interconnect to be a single twisted pair cable (shielded or unshielded) with  $120-\Omega$  characteristic impedance ( $Z_0$ ). Resistors equal to the characteristic impedance of the line should be used to terminate both ends of the cable to prevent signal reflections. Unterminated drop-lines (stubs) connecting nodes to the bus should be kept as short as possible to minimize signal reflections. The termination may be in a node, but if nodes may be removed from the bus, the termination must be carefully placed so that it is not removed from the bus.

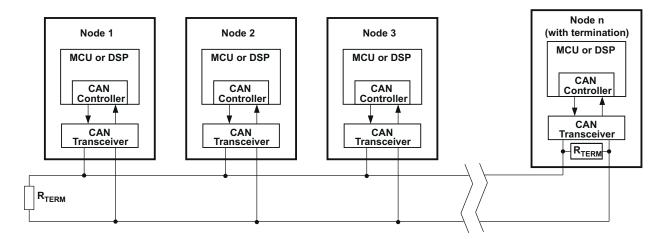


Figure 9-2. Typical CAN Bus

Termination may be a single  $120-\Omega$  resistor at the end of the bus, either on the cable or in a terminating node. If filtering and stabilization of the common mode voltage of the bus is desired, then split termination may be used. (See Figure 9-3). Split termination improves the electromagnetic emissions behavior of the network by eliminating fluctuations in the bus common-mode voltages at the start and end of message transmissions.



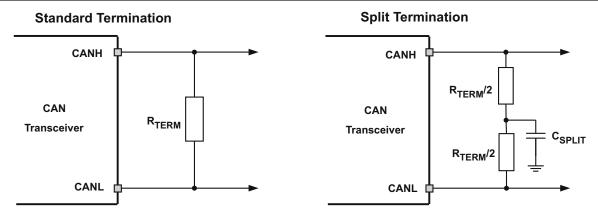


Figure 9-3. CAN Bus Termination Concepts

### 9.2.3 Application Curve

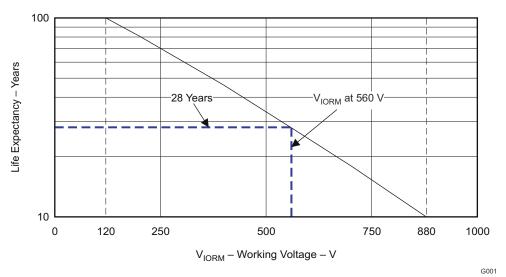


Figure 9-4. Life Expectancy vs Working Voltage (ISO1050DUB)

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# 10 Power Supply Recommendations

### 10.1 General Recommendations

To ensure reliable operation at all data rates and supply voltages, a 0.1- $\mu$ F bypass capacitor is recommended at input and output supply pins ( $V_{CC1}$  and  $V_{CC2}$ ). The capacitors should be placed as close to the supply pins as possible. If only a single primary-side power supply is available in an application, isolated power can be generated for the secondary-side using Texas Instruments' SN6505 and SN6501 based power supply solution. For such applications, detailed power supply design and transformer selection recommendations are available in SN6505 and SN6501 data sheets (SLLSEP9, SLLSEA0).



### 11 Layout

### 11.1 Layout Guidelines

A minimum of four layers is required to accomplish a low EMI PCB design (see Figure 11-1). Layer stacking should be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane and low-frequency signal layer.

- Routing the high-speed traces on the top layer avoids the use of vias (and the introduction of their inductances) and allows for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Placing a solid ground plane next to the high-speed signal layer establishes controlled impedance for transmission line interconnects and provides an excellent low-inductance path for the return current flow.
- Placing the power plane next to the ground plane creates additional high-frequency bypass capacitance of approximately 100 pF/in<sup>2</sup>.
- Routing the slower speed control signals on the bottom layer allows for greater flexibility as these signal links usually have margin to tolerate discontinuities such as vias.

If an additional supply voltage plane or signal layer is needed, add a second power / ground plane system to the stack to keep it symmetrical. This makes the stack mechanically stable and prevents it from warping. Also the power and ground plane of each power system can be placed closer together, thus increasing the high-frequency bypass capacitance significantly.

For detailed layout recommendations, see Application Note SLLA284, Digital Isolator Design Guide.

#### 11.1.1 PCB Material

For digital circuit boards operating below 150 Mbps, (or rise and fall times higher than 1 ns), and trace lengths of up to 10 inches, use standard FR-4 epoxy-glass as PCB material. FR-4 (Flame Retardant 4) meets the requirements of Underwriters Laboratories UL94-V0, and is preferred over cheaper alternatives due to its lower dielectric losses at high frequencies, less moisture absorption, greater strength and stiffness, and its self-extinguishing flammability-characteristics.

#### 11.2 Layout Example

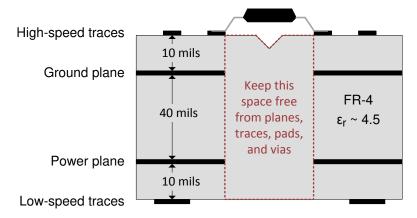


Figure 11-1. Recommended Layer Stack

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## 12 Device and Documentation Support

### 12.1 Documentation Support

#### 12.1.1 Related Documentation

- Texas Instruments, High-Voltage Lifetime of the ISO72x Family of Digital Isolatorsapplication report
- Texas Instruments, SN6505x Low-Noise 1-A Transformer Drivers for Isolated Power Supplies data sheet
- Texas Isntruments, Transformer Driver for Isolated Power Supplies data sheet
- Texas Instruments, Digital Isolator Design Guide application report
- Texas Instruments, Isolation Glossary application report

### 12.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Subscribe to updates* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 12.3 Support Resources

TI E2E<sup>™</sup> support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

#### 12.4 Trademarks

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### 12.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 12.6 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

### 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

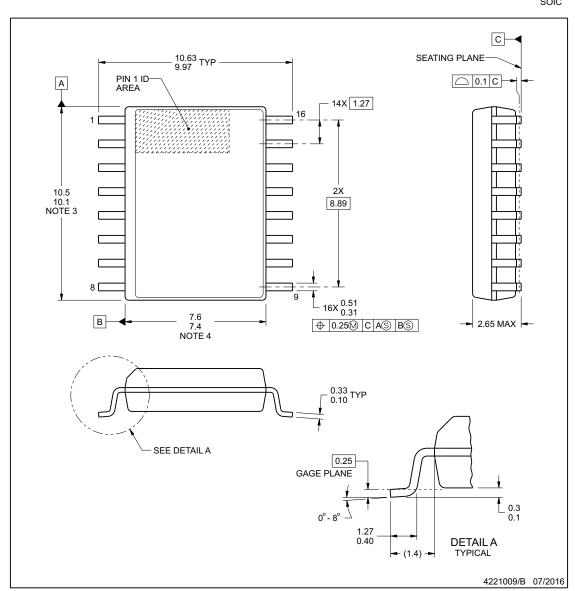




### **PACKAGE OUTLINE**

SOIC - 2.65 mm max height

# **DW0016B**



#### NOTES:

- 1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.

  4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
- 5. Reference JEDEC registration MS-013.

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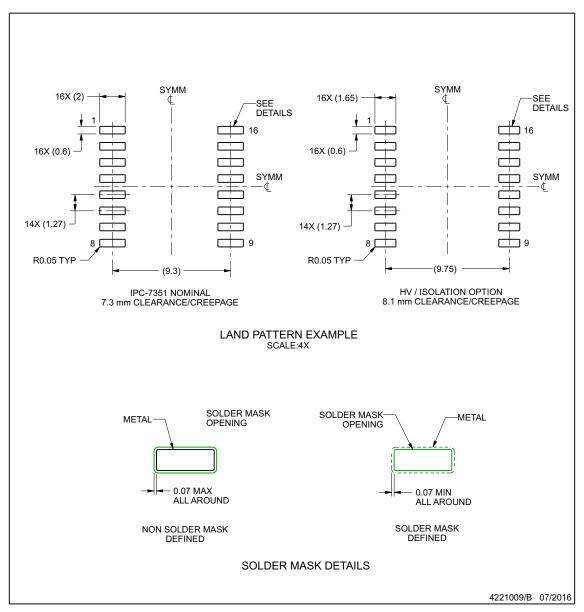


### **EXAMPLE BOARD LAYOUT**

# DW0016B

SOIC - 2.65 mm max height

SOIC



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

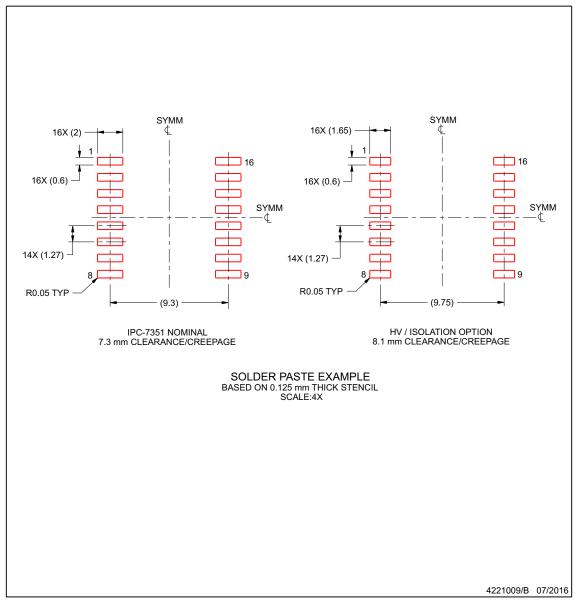
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### **EXAMPLE STENCIL DESIGN**

# **DW0016B**

SOIC - 2.65 mm max height



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

  9. Board assembly site may have different recommendations for stencil design.

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#### PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
							(6)				
ISO1050DUB	LIFEBUY	SOP	DUB	8	50	RoHS & Green	NIPDAU	Level-4-260C-72 HR	-55 to 105	ISO1050	
ISO1050DUBR	ACTIVE	SOP	DUB	8	350	RoHS & Green	NIPDAU	Level-4-260C-72 HR	-55 to 105	ISO1050	Samples
ISO1050DW	LIFEBUY	SOIC	DW	16	40	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-55 to 105	ISO1050	
ISO1050DWR	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-55 to 105	ISO1050	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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# **PACKAGE OPTION ADDENDUM**

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

# **PACKAGE MATERIALS INFORMATION**

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### TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISO1050DUBR	SOP	DUB	8	350	330.0	24.4	10.9	10.01	5.85	16.0	24.0	Q1
ISO1050DUBR	SOP	DUB	8	350	330.0	24.4	13.1	9.75	6.0	16.0	24.0	Q1
ISO1050DWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1



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### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISO1050DUBR	SOP	DUB	8	350	346.0	346.0	41.0
ISO1050DUBR	SOP	DUB	8	350	367.0	367.0	45.0
ISO1050DWR	SOIC	DW	16	2000	350.0	350.0	43.0

# **PACKAGE MATERIALS INFORMATION**

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### **TUBE**



#### \*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
ISO1050DUB	DUB	SOP	8	50	532.13	13	7300	6.6
ISO1050DUB	DUB	SOP	8	50	532.13	13.51	7.36	6.91
ISO1050DW	DW	SOIC	16	40	506.98	12.7	4826	6.6

7.5 x 10.3, 1.27 mm pitch

SMALL OUTLINE INTEGRATED CIRCUIT

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.





SOIC



#### NOTES:

- 1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing
- per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
- 5. Reference JEDEC registration MS-013.



SOIC



#### NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SOIC



#### NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.





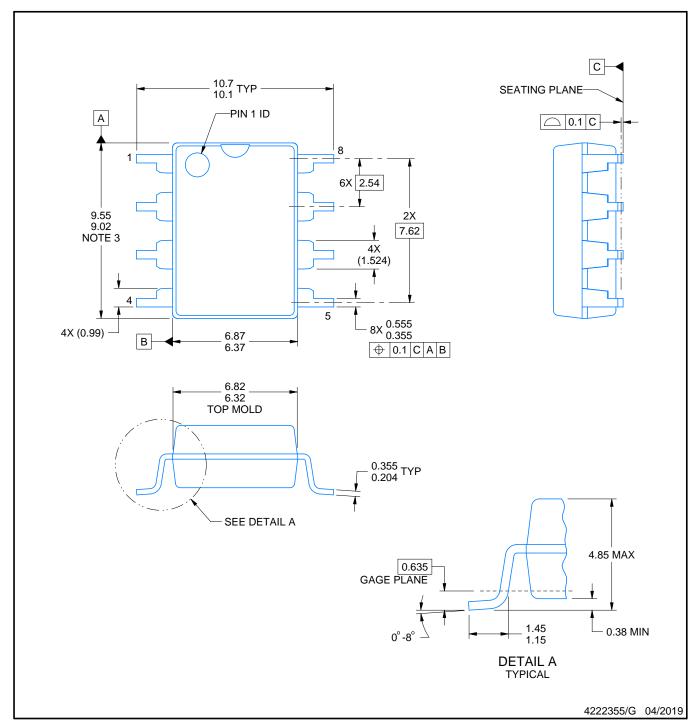
Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.

4207614/E





SMALL OUTLINE PACKAGE



#### NOTES:

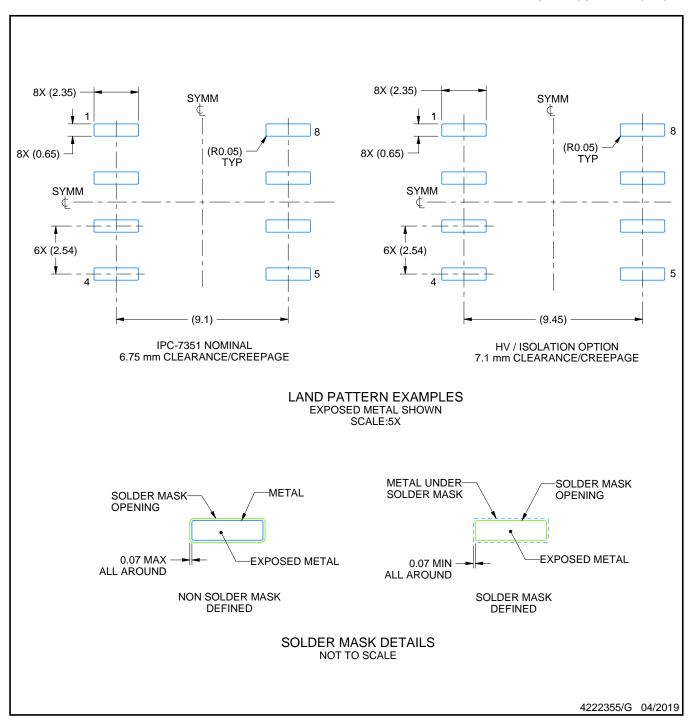
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  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.254 mm per side.



SMALL OUTLINE PACKAGE

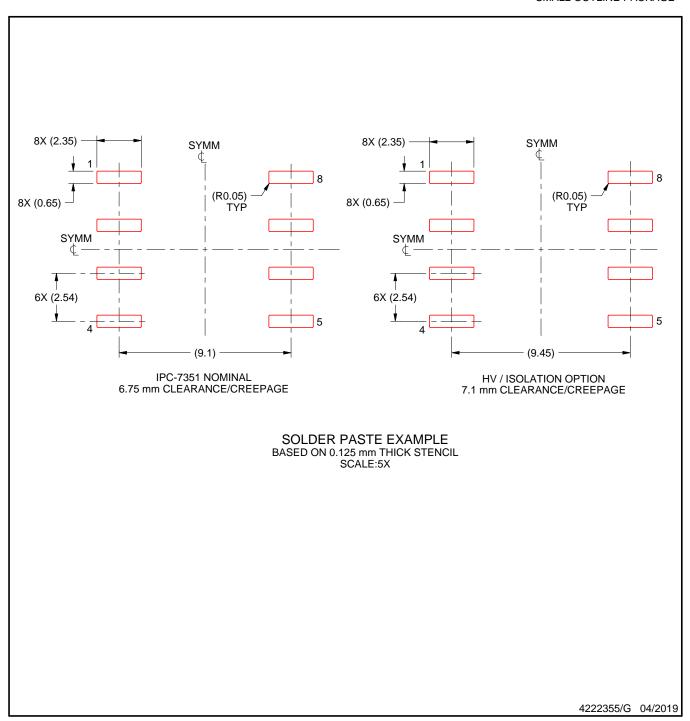


NOTES: (continued)

- 4. Publication IPC-7351 may have alternate designs.
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SMALL OUTLINE PACKAGE



NOTES: (continued)

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